

Series Resonant Converter

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ABSTRACT: Increasing the frequency of operation of power converters is desirable, as it allows the size of circuit magnetic and capacitors to be reduced, leading to cheaper and more compact circuits. However, by increasing the frequency of operation switching losses are increased hence reducing the system's efficiency. One solution to this problem is to replace the "chopper" switch with a "Resonant" switch, which uses the resonances of circuit capacitances and inductances to shape the waveform of either the current or the voltage across the switching element, such that when switching takes place, there is no current through or voltage across it, and hence no power dissipation. A circuit employing this technique is called resonant converter. Resonant converter is recent technology which is used in X-ray machine, Electrical vehicle charger etc. Resonant converter reduces switching loss through mechanism known as zero current switching and zero voltage switching. When the series resonant converter are operated below resonance the zero current phenomenon can occur and circuit acts as capacitive circuit, leading power factor. When resonant converter operate above resonance the zero voltage phenomenon can occurs and acts as inductive circuit, lagging power factor. Electrical Vehicle Battery Charger is an important part in automobile.

I. INTRODUCTION

Resonant converters operate at relatively high switching frequencies, and this enables the use of small inductive components which improve the dynamic behaviour and reduce the size of the converter. Despite the above-mentioned benefits of SMPS's, there are several parameters, which are not desired and have a strong influence on the converters behaviour being mainly:[1,2]

- Non-linear components in the converter structure,
- Line and load variations, and
- Electro-magnetic interferences (EMI).

The Resonant converter has non-linear components, the value of which changes non-linearly if the converter is disturbed or may change within time. These parameters force the converter to deviate from the desired operating condition. If the parameter deviation increases, this will cause the converter not to operate in steady state. Each control method has its own advantages and drawbacks due to which that particular control method appears to be the most suitable control method under specific conditions, compared to other control methods. It is always demanded to obtain a control Method that has the best performances under any conditions [4].

II. OPERATION OF SERIES RESONANT CONVERTER

In resonant converter chargers, drawbacks of switch mode chargers are overcome. Rectified AC input is turned into high-frequency AC using high frequency switching device (e.g. MOSFET). Voltage or zero current conditions are achieved using resonant tank circuit and a high-frequency transformer reduces the voltage to the exact level needed to charge the battery[6].

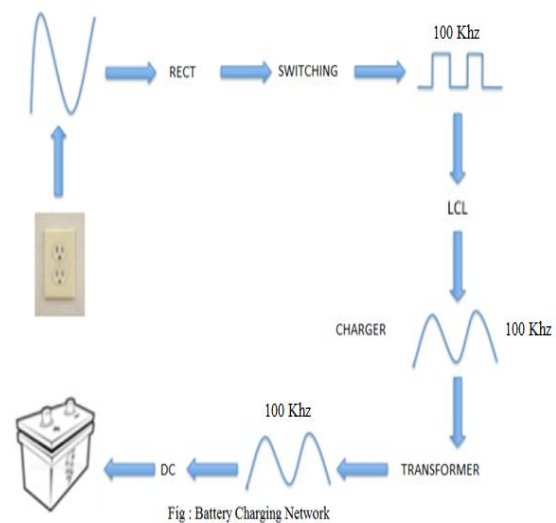


Fig.1 Block Diagram of Series Resonant Converter[5]

In the coming years, electric vehicles are predicted to dominate large segments of the automotive industry. Presently there are no widely accepted standards for electric vehicle charging techniques and many different methods are in use. The electric vehicle charger using a series L-C-L type resonant converter configuration will incorporate the benefits of maximized safety, efficiency, speed and ease of use. Fig.2 shown below represents battery charging circuit. In this circuit L-C-L type resonant converter configuration will be used and MOSFET will be used as a switch because it has low switching losses. It is proposed to design series resonant converter for charging of electric vehicle battery. The sequence of design is shown in figure below.[7].

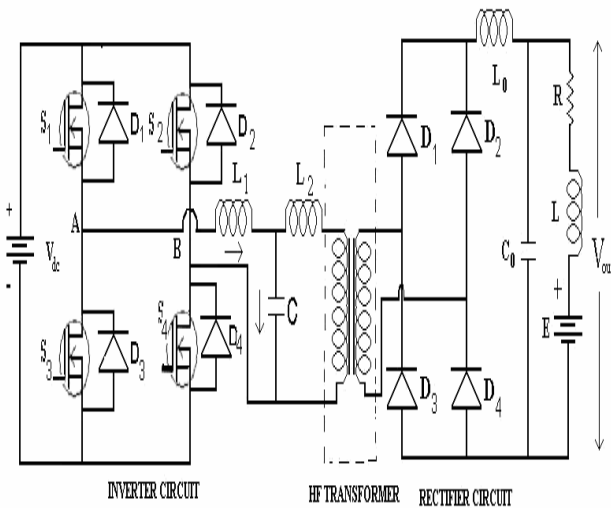


Fig.2 Circuit Diagram of Series Resonant Converter[7]

The aim of this project is to develop a fast and efficient method of electric vehicle battery charger using L-C-L type Resonant Converter topology. For ease of use, the entire system will be monitored by microcontroller which will control the power output of the charger and monitor the state of batteries to protect system from line born disturbances.

1. Zero –voltage switching:

When the series resonant converter is operated above resonance, the zero voltage switching phenomenon’s can occur, in which the circuit causes the transistor voltage to become zero before the controller turns the transistor on. With the minor circuit modification, the transistor turn-off transition can also be caused to occur at zero voltages. This process can lead the significant reductions in the switching losses of converters based on MOSFETs & Diodes.

For the full bridge circuit of fig13, the switch output voltage $V_s(t)$, & its fundamental component $V_{s1}(t)$, as well as the approximately sinusoidal tank current waveform $i_s(t)$, are plotted in fig18.

For the half cycle $0 < t < T_s/2$, the switch voltage $V_s(t)$ is equal to $+V_g$. For $0 < t < t_\alpha$, the current $i_s(t)$ is negative & diodes D_1 & D_4 conduct. Transistor Q_1 & Q_4 conduct when $i_s(t)$ is positive, over the interval $t_\alpha < t < T_s/2$. The waveform during $T_s/2 < t < T_s$ are symmetrical. Since the zero crossing of $V_s(t)$ leads the zero crossing of $i_s(t)$, the transistor conduct after their respective anti parallel diode.

In general, zero voltage switching can occurs. When the resonant tank present an effective inductive load to the switches, & hence the switch voltage zero crossing occur before the switch current zero crossing.

The transistor turn-off transition in fig. is similar to that of PWM switch. In converters the employ IGBTs or other minority-carrier devices, significant switching loss may occur at the turn-off transition. The current tailing phenomenon causes Q_1 to pass through a period of high instantaneous power dissipation & switching loss occurs. An additional advantage of zero-voltage switching is the reduction of EMI associated with device capacitance.

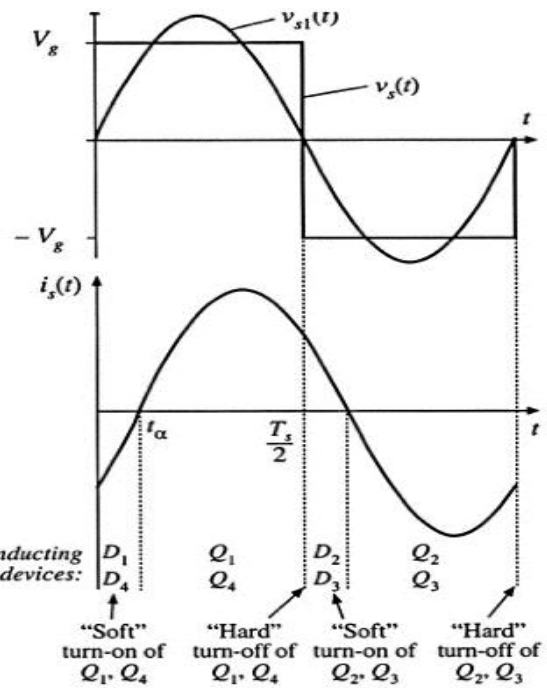


Fig.3 Switch Network Output Waveforms for Series Resonant Converter

III. SINUSOIDAL ANALYSIS OF RESONANT CONVERTERS

The class of resonant converters contains a controlled switch network (Ns) which drives a linear resonant tank network (Nr). In resonant inverter, the tank network drives a resistive load . The reactive component of the load impedances, if any, can be effectively incorporated into the tank network.

In the case of resonant DC-DC converter, the resonant tank network is connected to an uncontrolled rectifier network (Nr), filter network (Nf), load R. Many well known converters can be represented in this form, which includes the series, parallel, LCC topology.

In the most recommended modes of operation, the controlled switch network produces a square wave output voltage $V_s(t)$ whose frequency (F_s) is close to tank network’s resonant frequency (F_r). In the case where the resonant tank responds primarily to the fundamental component F_s of switch waveform of frequencies F_s . In the case where the negligible response at the harmonic frequencies $n(f_s)$, Where $n=3, 5, 7, \dots$ then the tank waveform’s are well approximated by their fundamental components.

As shown in the fig.9, this is indeed case where the tank network contains a high Q-resonance at or near the switching frequencies and a low pass characteristics at higher frequencies. Hence neglect harmonics and compute the relationship between the fundamental components of the tank terminal waveforms $V_s(t)$, $i_s(t)$, $i_r(t)$ and $V_r(t)$.

1. Controlled switch network model:

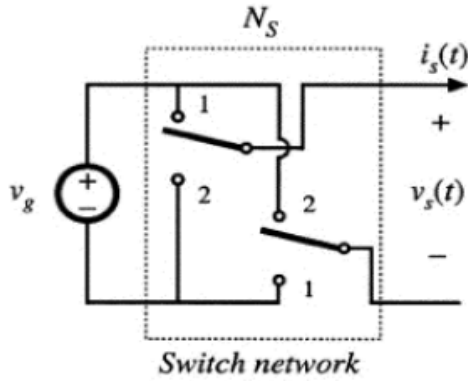


Fig.4 Ideal Switch Network

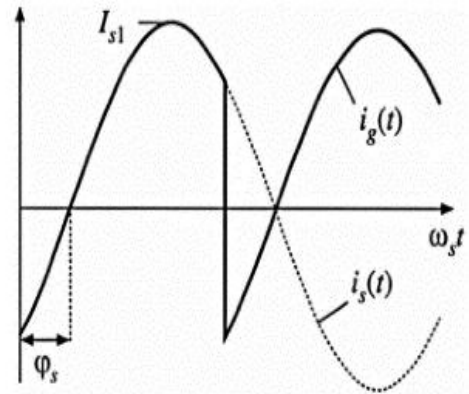


Fig.6 Switch Network Waveforms I_s (t) and I_r (t).

$$V_{s1} = -\frac{4V_g}{\pi} \sin(\omega_{st}) = V_{s1} \sin(\omega_{st})$$

$$V_s(t) = \frac{4V_g}{\pi} \sum_{n=1,3,5} \frac{1}{n} \sin(n \omega_s t)$$

$$i_{st}(t) \approx I_{s1} \sin(\omega_{st} - \phi)$$

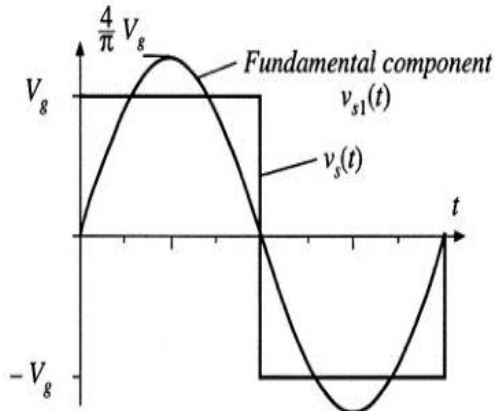


Fig.5 Switch Network Output Voltage V_s (t) and its Fundamental Component V_{s1}(t)

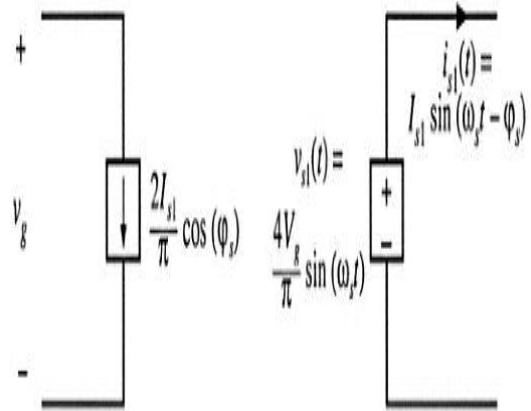


Fig.7 Equivalent Circuit For The Switch Network

$$\langle i_g(t) \rangle_{T_s} = \frac{2}{T_s} \int_0^{\frac{T_s}{2}} i_s(\tau) d\tau$$

$$= \frac{2}{T_s} \int_0^{\frac{T_s}{2}} I_{s1} \sin(\omega_s \tau - \phi_s) d\tau$$

$$= \frac{2}{\pi} I_{s1} \cos(\phi_s)$$

IV. CONCLUSION

In order to study and analyze the Switch Mode Control effect on the converter behavior and to compare it to another control method, a comparison will be done for the output voltage and the inductor current responses of the Series LCL

T- type Resonant converter controlled with the PID control and the SMC. Test results will be obtained in steady state and under dynamic conditions.

Resonant converters have advantages such as reduction in cost, size and weight of the power supply. Fast transient response, reduction in switching losses, di/dt and dv/dt stresses of power devices gives additional advantages of using resonant converters. It also reduces Electro-Magnetic Interference (EMI).

Resonant converter is recent technology which is used in X-ray machine, high HV pulsed load applications, Ozone generation using LCC RC, high-power industrial application of CO₂ laser also needs HV power supply, Electrical vehicle charger etc.

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